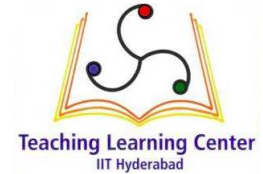




## Gate problems in DSP



**Abstract :** *These problems have been selected from GATE question papers and can be used for conducting tutorials in courses related to the course Digital Signal Processing in practice.*

1) If the impulse response of a discrete-time system is  $h[n] = -5^n u[-n - 1]$ , then the system function  $H(z)$  is equal to

- (A)  $\frac{-z}{z-5}$  and the system is stable  
 (B)  $\frac{z}{z-5}$  and the system is stable  
 (C)  $\frac{-z}{z-5}$  and the system is unstable  
 (D)  $\frac{z}{z-5}$  and the system is unstable

2) A sequence  $x(n)$  with the z-transform  $X(z) = z^4 + z^2 - 2z + 2 - 3z^{-4}$  is applied as an input to a linear, time-invariant system with the impulse response  $h(n) = 2\delta(n - 3)$  where

$$\delta(n) = \begin{cases} 1, & \text{if } n=0 \\ 0, & \text{otherwise} \end{cases}$$

The output at  $n=4$  is

- (A) -6 (B) 0 (C) 2 (D) -4

Data for **Q.3-4** are given below. Solve the problems and choose the correct answers.

The system under consideration is an RC low-pass filter (RC-LPF) with  $R=1.0\text{K}\Omega$  and  $C=1.0\ \mu\text{F}$

3) Let  $H(f)$  denote the frequency response of the RC-LPF. Let  $f_1$  be the highest frequency such

that  $0 \leq |f| \leq f_1, \frac{|H(f_1)|}{H(0)} \geq 0.95$ . Then  $f_1$  (in HZ) is

- (A) 327.8 (B) 163.9 (C) 52.2 (D) 104.4

4) The impulse response  $h[n]$  of a linear time-invariant system is given by  $h[n] = u[n + 3] + u[n - 2] - 2u[n - 7]$  where  $u[n]$  is the unit step sequence. The above system is

- (A) Stable but not causal  
 (B) Stable and Causal  
 (C) Causal but unstable  
 (D) Unstable and not Causal

5) The z-transform of a system is  $H(z) = \frac{z}{z-0.2}$ . If the ROC is  $|z| < 0.2$ , then the impulse response of the system is

- (A)  $(0.2)^n u[n]$  (C)  $-(0.2)^n u[n]$   
 (B)  $(0.2)^n u[-n - 1]$  (D)  $-(0.2)^n u[-n - 1]$

6) consider the sequence  $x[n] = [4 - j5 \ 1 + j2 \ 4]$  The conjugate anti-symmetric part of the sequence is

- (A)  $[-4 - j2.5 \ j2 \ 4 - j2.5]$   
 (B)  $[-j2.5 \ 1 \ j2.5]$

(C)  $[-j5 \quad j2 \quad 0]$

(D)  $[-4 \quad 1 \quad 4]$

7) A causal LTI system is described by the difference equation  $2y[n] = ay[n-2] - 2x[n] + bx[n-1]$

the system is stable only if

(A)  $|a| = 2, |b| < 2$

(B)  $|a| > 2, |b| > 2$

(C)  $|a| < 2, \text{any value of } b$

(D)  $|b| < 2, \text{any value of } a$

8) The impulse response  $h[n]$  of a linear time invariant system is given as

$$h[n] = \begin{cases} -2\sqrt{2}, & \text{if } n=1, -1 \\ 4\sqrt{2}, & n=2, -2 \\ 0, & \text{otherwise.} \end{cases}$$

If the input to the above system is the sequence  $e^{\frac{jpn}{4}}$ , then the output is

(A)  $4\sqrt{2}e^{\frac{jpn}{4}}$

(C)  $4e^{\frac{jpn}{4}}$

(B)  $4\sqrt{2}e^{-\frac{jpn}{4}}$

(D)  $-4e^{\frac{jpn}{4}}$

9) The region of convergence of Z-transform of the sequence  $(\frac{5}{6})^n u(n) - (\frac{6}{5})^n u(-n-1)$  must be

(A)  $|z| < \frac{5}{6}$

(C)  $\frac{5}{6} < |z| < \frac{5}{6}$

(B)  $|z| > \frac{6}{5}$

(D)  $\frac{6}{5} < |z| < \infty$

10) A signal  $x(n) = \sin(\omega_0 n + \phi)$  is the input to a LTI system frequency response  $H(e^{j\omega})$ . If the output of the system is  $Ax(n - n_0)$ , then the most general form of  $\angle H(e^{j\omega})$  will be

(A)  $-n_0\omega_0 + \beta$  for any arbitrary real  $\beta$

(B)  $-n_0\omega_0 + 2\pi k$  for any arbitrary integer  $k$ .

(C)  $n_0\omega_0 + 2\pi k$  for any arbitrary integer  $k$ .

(D)  $-n_0\omega_0 + \phi$

11) A system with input  $x[n]$  and the output  $y[n]$  is given as  $y[n] = (\sin\frac{5}{6}\pi n)x[n]$ . The system is

(A) Linear, stable and invertible

(B) non-linear, stable and non-invertible

(C) linear, stable and non-invertible

(D) linear, unstable and invertible

12) A 5-point sequence  $x[n]$  is given as

$x[-3] = 1, x[-2] = 1, x[-1] = 0, x[0] = 5, x[1] = 1$ . Let  $X(e^{j\omega})$  denote the discrete-time Fourier transform of  $x[n]$ . The value of  $\int_{-\pi}^{\pi} X(e^{j\omega}) d\omega$  is :

(A) 5

(B)  $10\pi$

(C)  $16\pi$

(D)  $5 + j10\pi$

13) The z-transform  $X[z]$  of a sequence  $x[n]$  is given by  $X[z] = \frac{0.5}{1-2z^{-1}}$ . It is given that the region of convergence of  $X[z]$  includes the unit circle. The value of  $x[0]$  is:

(A) -0.5

(B) 0

(C) 0.25

(D) 0.5

14) A discrete time linear shift-invariant system has an impulse response  $h[n]$  with  $h[0] = 1, h[1] = -1, h[2] = -2$  and zero otherwise. The system is given an input sequence  $x[n]$  with  $x[0] = x[2] = 1$ , and zero otherwise. The number of nonzero samples in the output sequence  $y[n]$ , and the value of  $y[2]$  are, respectively

(A) 5,2

(C) 6,1

(B)  $\frac{5z}{z-e^{-0.05}}, |z| < e^{-0.05}$

(B) 6,2

(D) 5,3

(C)  $\frac{5z}{z-e^{-5}}, |z| > e^{-0.05}$

- 15)  $\{x(n)\}$  is real-valued periodic sequence with a period N.  $x(n)$  and  $X(k)$  form N-point Discrete Fourier Transform (DFT) pairs. The DFT  $Y(k)$  of the sequence  $y(n) = \frac{1}{N} \sum_{r=0}^{N-1} x(r)x(n+r)$

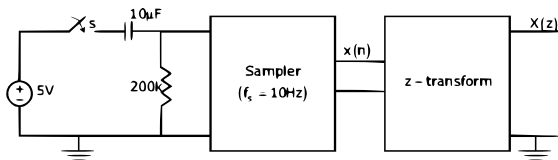
(A)  $|X(k)|^2$

(B)  $\frac{1}{N} \sum_{r=0}^{N-1} X(r) * X(k+r)$

(C)  $\frac{1}{N} \sum_{r=0}^{N-1} X(r)X(k+r)$

(D) 0

**DATA FOR Q.16 AND 17** In the following network, the switch is closed at  $t=0^-$  and the sampling starts from  $t=0$ . The sampling frequency is 10Hz.



- 16) The samples  $x(n)$  at  $n=0,1,2,\dots$  given by

(A)  $5(1 - e^{-0.05n})$

(B)  $5e^{-0.05n}$

(C)  $5(1 - e^{-5n})$

(D)  $5e^{-5n}$

- 17) The expression and the region of convergence of the z-transform of the sampled signal are

(A)  $\frac{5z}{z-e^{-5}}, |z| < e^{-5}$

- 18) The ROC of Z-transform of the discrete time sequence  $x(n) = (\frac{1}{3})^n u(n) - (\frac{1}{2})^n u(-n-1)$  is

(A)  $\frac{1}{3} < |z| < \frac{1}{2}$

(C)  $|z| < \frac{1}{3}$

(B)  $|z| > \frac{1}{2}$

(D)  $2 < |z| < 3$

- 19) A system with transfer function  $H(z)$  has impulse response  $h(x)$  defined as  $h(2)=1, h(3)=-1$  and  $h(k)=0$  otherwise. Consider the following statements.

S1:  $H(z)$  is a low pass filter

S2:  $H(z)$  is a FIR filter

which of the following is correct ?

(A) Only S2 is true

(B) Both S1 and S2 are false.

(C) Both S1 and S2 are true, and S2 is a reason for S1

(D) Both S1 and S2 are true, but S2 is not a reason for S1

- 20) Consider the z-transform  $X(z) = 5z^2 + 4z^{-1} + 3; 0 < |z| < \infty$ . The inverse z-transform  $x[n]$

(A)  $5\delta[n+2] + 3\delta[n] + 4\delta[n-1]$

(B)  $5\delta[n-2] + 3\delta[n] + 4\delta[n+1]$

(C)  $5u[n+2] + 3u[n] + 4u[n-1]$

(D)  $5u[n-2] + 3u[n] + 4u[n+1]$

- 21) Two discrete time systems with impulse responses  $h_1[n] = \delta[n-1]$  and  $h_2[n] = \delta[n-2]$

are cascade. The overall impulse response of the cascaded system is

- (A)  $\delta[n - 1] + \delta[n - 2] + \delta[n - 3]$   
 (B)  $\delta[n - 4]$  (C)  $\delta[n - 3]$   
 (D)  $\delta[n - 1]\delta[n - 2]$

- 22) The transfer function of a discrete time LTI system is given by  $H(z) = \frac{2 - \frac{3}{4}z^{-1}}{1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2}}$

Consider the following statements:

**S1:**The system is stable and causal for ROC:  $|z| > \frac{1}{2}$

**S2:**The system is stable but not causal for ROC:  $|z| < \frac{1}{4}$

**S3:**The system is neither stable nor causal for ROC:  $\frac{1}{4} < |z| < \frac{1}{2}$

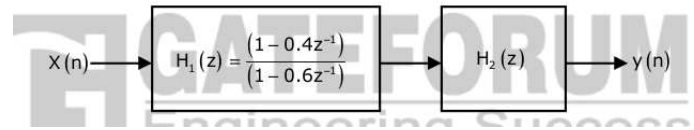
Which one of the following statements is valid?

- (A) Both S1 and S2 are true.  
 (B) Both S2 and S3 are true.  
 (C) Both S1 and S3 are true.  
 (D) S1, S2 and S3 are all true.

- 23) A system is defined by its impulse response  $h(n) = 2^n u(n - 2)$ . The system is

- (A) stable and causal  
 (B) causal but not stable  
 (C) stable but not causal  
 (D) unstable and non-causal

- 24) Two systems  $H_1(z)$  and  $H_2(z)$  are connected in cascade as shown below. The overall output  $y(n)$  is the same as the input  $x(n)$  with a one unit delay. The transfer function of the second system  $H_2(z)$  is



- (A)  $\frac{(1-0.6z^{-1})}{z^{-1}(1-0.4z^{-1})}$  (C)  $\frac{z^{-1}(1-0.4z^{-1})}{(1-0.6z^{-1})}$   
 (B)  $\frac{z^{-1}(1-0.6z^{-1})}{(1-0.4z^{-1})}$  (D)  $\frac{(1-0.4z^{-1})}{z^{-1}(1-0.6z^{-1})}$

- 25) The first 6 points of the 8-point DFT of a real valued sequence are  $5, 1-j3, 0, 3-j4, 0$  and  $3+j4$ . The last two points of the DFT are respectively

- (A)  $0, 1-j3$  (C)  $1+j3, 5$   
 (B)  $0, 1+j3$  (D)  $1-j3, 5$

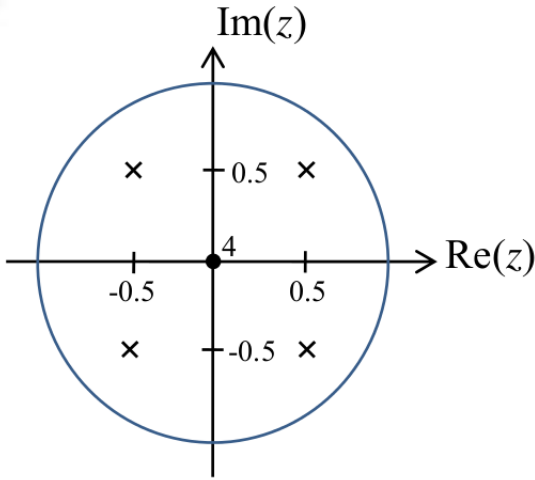
- 26) Let  $x[n]=x[-n]$ . Let  $X(z)$  be the z-transform of  $x[n]$ . If  $0.5+j0.25$  is a zero of  $X(z)$  then one of the following must be a zero of  $X(z)$ .

- (A)  $0.5 - j0.25$  (C)  $\frac{1}{0.5-j0.25}$   
 (B)  $\frac{1}{0.5+j0.25}$  (D)  $2 + j4$

- 27) The input-output relationship of a causal stable LTI system is given as  $y[n] = \alpha y[n-1] + \beta x[n]$ . If the impulse response  $h[n]$  of this system satisfies the condition  $\sum_{n=0}^{\infty} h[n] = 2$ , the relationship between  $\alpha$  and  $\beta$  is

- (A)  $\alpha = 1 - \frac{\beta}{2}$  (C)  $\alpha = 2\beta$   
 (B)  $\alpha = 1 + \frac{\beta}{2}$  (D)  $\alpha = -2\beta$

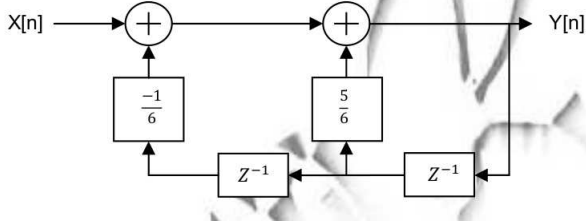
- 28) The pole-zero diagram of causal and stable discrete-time system is shown in figure. The zero at the origin has multiplicity 4. The impulse response of the system is  $h[n]$ . If  $h[0] = 1$ , we can conclude



- (A)  $h[n]$  is real for all  $n$
- (B)  $h[n]$  is purely imaginary for all  $n$
- (C)  $h[n]$  is real for only even  $n$
- (D)  $h[n]$  is purely imaginary for only odd  $n$

29) Consider the signal  $x[n] = 6\delta[n + 2] + 3\delta[n + 1] + 8\delta[n] + 7\delta[n - 1] + 4\delta[n - 2]$ . If  $X(e^{j\omega})$  is the discrete-time Fourier transform of  $x[n]$ . Then  $\frac{1}{\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) \sin^2(2\omega) d\omega$  is equal to \_\_\_\_\_

30) For the discrete-time shown in the figure, the poles of the system transfer function are located at



- (A) 2, 3
- (B)  $\frac{1}{2}, 3$
- (C)  $\frac{1}{2}, \frac{1}{3}$
- (D) 2,  $\frac{1}{3}$

31) The DFT of vector  $[a \ b \ c \ d]$  is the vector  $[\alpha \ \beta \ \gamma \ \delta]$ . Consider the product

$$[p \ q \ r \ s] = [a \ b \ c \ d] \begin{bmatrix} a & b & c & d \\ d & a & b & c \\ c & d & a & b \\ b & c & d & a \end{bmatrix}$$

The DFT of the vector  $[p \ q \ r \ s]$  is a scaled version of

- (A)  $[\alpha^2 \ \beta^2 \ \gamma^2 \ \delta^2]$
- (B)  $[\sqrt{\alpha} \ \sqrt{\beta} \ \sqrt{\gamma} \ \sqrt{\delta}]$
- (C)  $[\alpha + \beta \ \beta + \delta \ \gamma + \delta \ \gamma + \alpha]$
- (D)  $[\alpha \ \beta \ \gamma \ \delta]$

32) Two sequences  $[a \ b \ c]$  and  $[A \ B \ C]$  are related as

$$\begin{bmatrix} A \\ B \\ C \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 \\ 1 & W_3^{-1} & W_3^{-2} \\ 1 & W_3^{-2} & W_3^{-4} \end{bmatrix} \begin{bmatrix} a \\ b \\ c \end{bmatrix} \quad \text{Where}$$

$$W_3 = e^{j\frac{2\pi}{3}}$$

If another sequence  $[p \ q \ r]$  is derived as

$$\begin{bmatrix} p \\ q \\ r \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 \\ 1 & W_3^1 & W_3^2 \\ 1 & W_3^2 & W_3^4 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \\ 0 & W_3^2 & 0 \\ 0 & W_3^4 & 0 \end{bmatrix} \begin{bmatrix} \frac{A}{3} \\ \frac{B}{3} \\ \frac{C}{3} \end{bmatrix}$$

Then the relationship between the sequences  $[p \ q \ r]$  and  $[a \ b \ c]$

- a)  $[p \ q \ r] = [b \ a \ c]$
- b)  $[p \ q \ r] = [b \ c \ a]$
- c)  $[p \ q \ r] = [c \ a \ b]$
- d)  $[p \ q \ r] = [c \ b \ a]$

33) Let  $h[n]$  be the impulse response of a discrete-time linear time invariant (LTI) filter. The impulse response is given by  $h[0] = \frac{1}{3}; h[1] = \frac{1}{3}; h[2] = \frac{1}{3}$  and  $h[n] = 0$  for  $n < 0$  and  $n > 2$ . Let  $H(\omega)$  be the discrete-time Fourier transform (DTFT) of  $h[n]$ , where  $\omega$  is the normalized angular frequency in radians. Given that

$H(\omega_0) = 0$  and  $0 < \omega < \pi$ , the value of  $\omega_0$  (in radians) is equal to \_\_\_\_\_

34) A discrete-time signal  $x[n] = \delta[n-3] + 2\delta[n-5]$  has  $z$ -transform  $X(z)$ . If  $Y(z) = X(-z)$  is the  $z$ -transform of another signal  $y[n]$ , then

(A)  $y[n] = x[n]$       (C)  $y[n] = -x[n]$

(B)  $y[n] = x[-n]$       (D)  $y[n] = -x[-n]$

35) The 4-point Discrete Fourier Transform (DFT) of a discrete time sequence 1,0,2,3 is

(A)  $[0, -2 + 2j, 2, -2 - 2j]$

(B)  $[2, 2 + 2j, 6, -2 - 2j]$

(C)  $[6, 1 - 3j, 2, 1 + 3j]$

(D)  $[6, -1 + 3j, 0, -1 - 3j]$